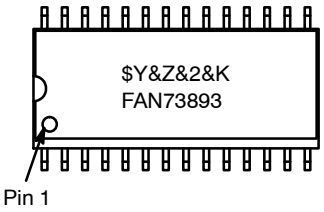
3-Phase Half-Bridge Gate-Drive IC

FAN73893



SOIC−28, 300 mils
CASE 751BM−01

MARKING DIAGRAM



FAN73893 = Specific Device Code
$Y = Logo
&Z = Assembly Plant Code
&2 = 2−Digit Date Code Format
&K = 2−Digits Lot Run Traceability Code

ORDERING INFORMATION

See detailed ordering and shipping information on page 15 of this data sheet.

Description

The FAN73893 is a monolithic three−phase half−bridge gate−drive IC designed for high−voltage, high−speed, driving MOSFETs and IGBTs operating up to +600 V.

onsemi's high−voltage process and common−mode noise−canceling technique provide stable operation of high−side drivers under high−dV_s/dt noise circumstances.

An advanced level−shift circuit allows high−side gate driver operation up to V_S = −9.8 V (typical) for V_{BS} = 15 V.

The protection functions include under−voltage lockout, inter−lock function and inverter over−current trip with an automatic fault−clear function. Over−current protection that terminates all six outputs can be derived from an external current−sense resistor. An open−drain fault signal is provided to indicate that an over−current or under−voltage shutdown has occurred. The UVLO circuits prevent malfunction when V_{DD} and V_{BS} are lower than the threshold voltage.

Output drivers typically source and sink 350 mA and 650 mA, respectively; which is suitable for three−phase half−bridge applications in motor drive systems.

Features

- Floating Channel for Bootstrap Operation to +600 V
- Typically 350 mA/650 mA Sourcing/Sinking Current−Driving Capability for All Channels
- Extended Allowable Negative V_S Swing to −9.8 V for Signal Propagation at V_{DD} = V_{BS} = 15 V
- Outputs Out of Phase with Input Signals
- Over−Current Shutdown Turns Off All Six Drivers
- Matched Propagation Delay for All Channels
- 3.3 V and 5.0 V Input Logic Compatible
- Adjustable Fault−Clear Timing
- Signal Interlocking of Every Phase to Prevent Cross−Conduction
- Common−Mode dV_s/dt Noise−Canceling Circuit
- Built−in Advanced Input Filter
- Built−in Soft Turn−Off Function
- Built−in UVLO Functions for All Channels
- This is a Pb−Free Device

Applications

- 3−Phase Motor Inverter Driver
- Air Conditioner, Washing Machine, Refrigerator, Dish Washer
- Industrial Inverter – Sewing Machine, Power Tool
- General−Purpose Three−Phase Inverter

Table 1. COMPARISION TABLE

Part	FAN73893MX	FAN73894MX	FAN73895MX	FAN73896MX
INPUT Type	Inverted	Inverted	Non−inverted	Non−inverted
V_{DDUV+} / V_{BSUV+} (Min / Typ / Max)	7.5 / 8.5 / 9.3 [V]	10.2 / 11.2 / 12 [V]	7.5 / 8.5 / 9.3 [V]	10.2 / 11.2 / 12 [V]
V_{DDUV-} / V_{BSUV-} (Min / Typ / Max)	7 / 8 / 8.7 [V]	9.7 / 10.7 / 11.4 [V]	7 / 8 / 8.7 [V]	9.7 / 10.7 / 11.4 [V]
Note	(Replacement for FAN73892MX)	−	(Replacement for FAN7389MX1)	−

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TYPICAL APPLICATION DIAGRAM

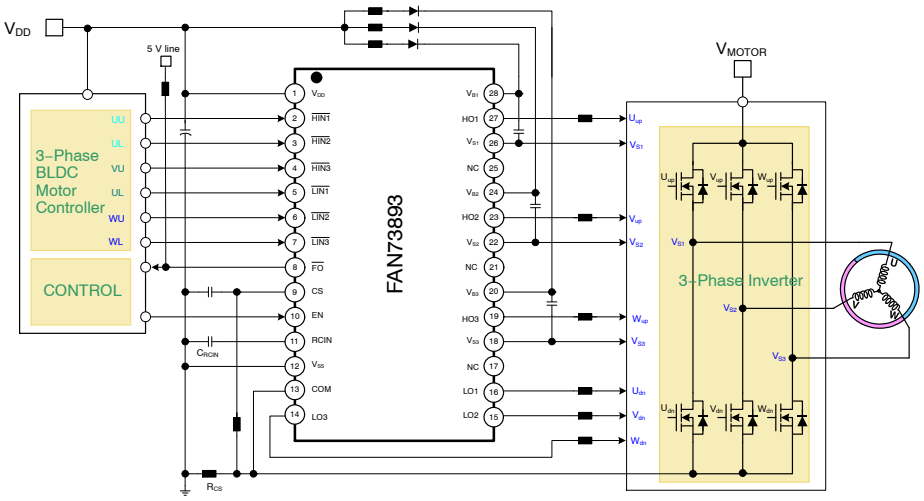


Figure 1. 3−Phase BLDC Motor Drive Application

INTERNAL BLOCK DIAGRAM

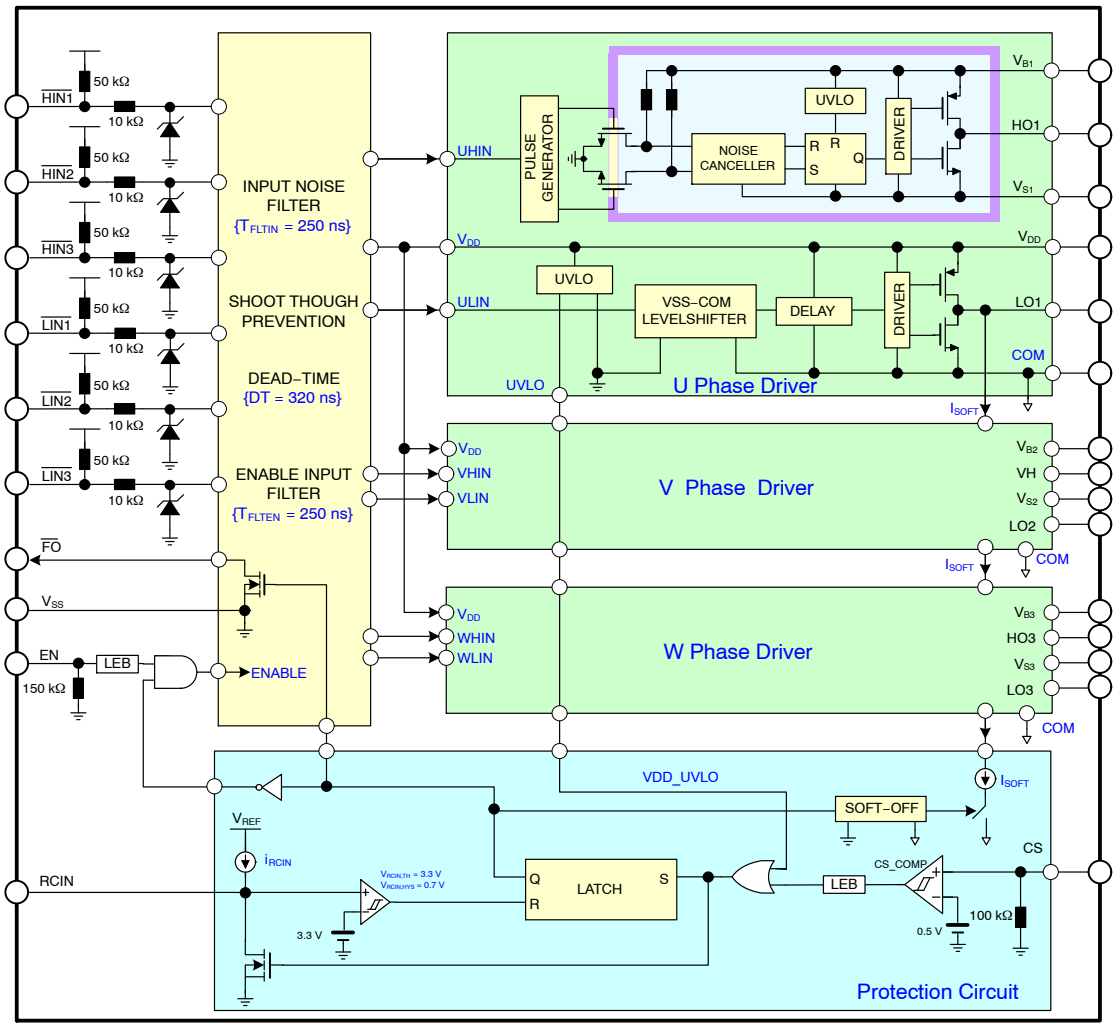


Figure 2. Functional Block Diagram

PIN CONFIGURATION

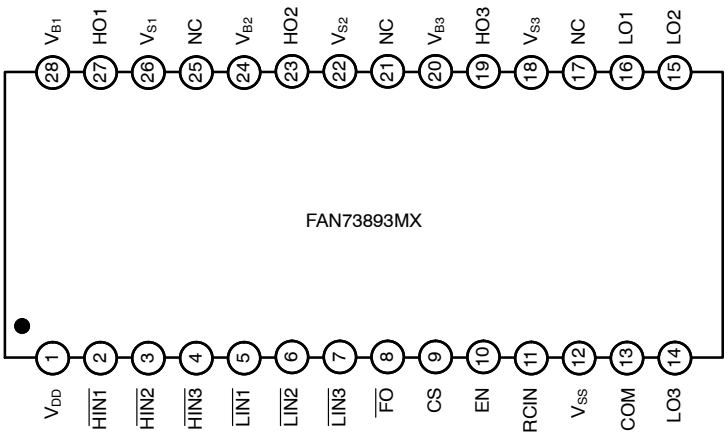


Figure 3. Pin Assignments

PIN DEFINITIONS

Pin	Symbol	Description
1	V_{DD}	Logic and low−side gate driver power supply voltage
2	$\overline{HIN1}$	Logic Input 1 for high−side gate 1 driver
3	$\overline{HIN2}$	Logic Input 2 for high−side gate 2 driver
4	$\overline{HIN3}$	Logic Input 3 for high−side gate 3 driver
5	$\overline{LIN1}$	Logic Input 1 for low−side gate 1 driver
6	$\overline{LIN2}$	Logic Input 2 for low−side gate 2 driver
7	$\overline{LIN3}$	Logic Input 3 for low−side gate 3 driver
8	$\overline{FO}$	Fault output with open drain (indicates over−current and low−side under−voltage)
9	CS	Analog input for over−current shutdown
10	EN	Logic input for shutdown functionality
11	RCIN	An external RC network input used to define the fault−clear delay
12	V_{SS}	Logic ground
13	COM	Low−side driver return
14	LO3	Low−side gate driver 3 output
15	LO2	Low−side gate driver 2 output
16	LO1	Low−side gate driver 1 output
17, 21, 25	NC	No connect
18	V_{S3}	High−side driver 3 floating supply offset voltage
19	HO3	High−side driver 3 gate driver output
20	V_{B3}	High−side driver 3 floating supply
22	V_{S2}	High−side driver 2 floating supply offset voltage
23	HO2	High−side driver 2 gate driver output
24	V_{B2}	High−side driver 2 floating supply
26	V_{S1}	High−side driver 1 floating supply offset voltage
27	HO1	High−side driver 1 gate driver output
28	V_{B1}	High−side driver 1 floating supply

ABSOLUTE MAXIMUM RATINGS (T_A = 25°C unless otherwise specified)

Symbol	Parameter	Min	Max	Unit
V_S	High−Side Floating Offset Voltage	$V_{B1,2,3}$ − 25	$V_{B1,2,3}$ + 0.3	V
V_B	High−Side Floating Supply Voltage	−0.3	625.0	V
V_{DD}	Low−Side and Logic−Fixed supply voltage	−0.3	25.0	V
V_{HO}	High−Side Floating Output Voltage $V_{HO1,2,3}$	$V_{S1,2,3}$ − 0.3	$V_{B1,2,3}$ + 0.3	V
V_{LO}	Low−Side Floating Output Voltage $V_{LO1,2,3}$	−0.3	V_{DD} + 0.3	V
V_{IN}	Input Voltage ($\overline{HINx}$, $\overline{LINx}$, CS, and EN) (Note 1)	V_{SS} − 0.3	V_{SS} + 5.5	V
V_{FO}	Fault Output Voltage ($\overline{FO}$)	−0.3	V_{DD} + 0.3	V
dV_S/dt	Allowable Offset Voltage Slew Rate	−	±50	V/ns
P_D	Power Dissipation (Note 2, 3)	−	1.4	W
θ_{JA}	Thermal Resistance	−	70	°C/W
T_J	Junction Temperature	−	150	°C
T_{STG}	Storage Temperature	−55	150	°C

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

1. All input voltage ($\overline{HINx}$, $\overline{LINx}$, CS, and EN) are referenced to V_{SS} and do not exceed maximum voltage rating.
2. Mounted on 76.2 x 114.3 x 1.6 mm PCB (FR−4 glass epoxy material). *Refer to the following standards:*
   JESD51−2: Integral circuit's thermal test method environmental conditions, natural convection;
   JESD51−3: Low effective thermal conductivity test board for leaded surface−mount packages.
3. Do not exceed maximum power dissipation (P_D) under any circumstances.

RECOMMENDED OPERATING CONDITIONS

Symbol	Parameter	Min	Max	Unit
$V_{B1,2,3}$	High−Side Floating Supply Voltage	$V_{S1,2,3}$ + 10	$V_{S1,2,3}$ + 20	V
$V_{S1,2,3}$	High−Side Floating Supply Offset Voltage	6 − V_{DD}	600	V
V_{DD}	Low−Side and Logic Fixed Supply Voltage	10	20	V
$V_{HO1,2,3}$	High−Side Output Voltage	$V_{S1,2,3}$	$V_{B1,2,3}$	V
$V_{LO1,2,3}$	Low−Side Output Voltage	COM	V_{DD}	V
V_{FO}	Fault Output Voltage ($\overline{FO}$)	V_{SS}	V_{DD}	V
V_{CS}	Current−Sense Pin Input Voltage	V_{SS}	V_{SS} + 5	V
V_{IN}	Logic Input Voltage ($\overline{HIN1,2,3}$ and $\overline{LIN1,2,3}$)	V_{SS}	V_{SS} + 5	V
V_{SS}	Logic Ground	−5	5	V
T_A	Ambient Temperature	−40	+125	°C

Functional operation above the stresses listed in the Recommended Operating Ranges is not implied. Extended exposure to stresses beyond the Recommended Operating Ranges limits may affect device reliability.

ELECTRICAL CHARACTERISTICS (V_{BIAS} (V_{DD}, $V_{BS1,2,3}$) = 15.0 V and T_A = 25°C unless otherwise specified. The V_{IN} and I_{IN} parameters are referenced to V_{SS} and are applicable to all six channels. The V_O and I_O parameters are referenced to $V_{S1,2,3}$ and COM and are applicable to the respective output leads: HO1,2,3 and LO1,2,3. The V_{DDUV} parameters are referenced to V_{SS}. The V_{BSUV} parameters are referenced to $V_{S1,2,3}$.)

Symbol	Parameter	Condition	Min	Typ	Max	Unit
LOW−SIDE POWER SUPPLY SECTION						
I_{QDD}	Quiescent V_{DD} Supply Current	$V_{LIN1,2,3}$ = 5 V or open, EN = 0 V	−	250	400	μA
I_{PDD}	Operating V_{DD} Supply Current	$f_{LIN1,2,3}$ = 20 kHz, rms Value	−	550	750	μA
V_{DDUV+}	V_{DD} Supply Under−Voltage Positive−Going Threshold	V_{DD} = Sweep	7.5	8.5	9.3	V
$V_{DDUV−}$	V_{DD} Supply Under−Voltage Negative−Going Threshold	V_{DD} = Sweep	7.0	8.0	8.7	V
V_{DDHYS}	V_{DD} Supply Under−Voltage Lockout Hysteresis	V_{DD} = Sweep	−	0.5	−	V
BOOTSTRAPPED POWER SUPPLY SECTION						
V_{BSUV+}	V_{BS} Supply Under−Voltage Positive−Going Threshold	$V_{BS1,2,3}$ = Sweep	7.5	8.5	9.3	V
$V_{BSUV−}$	V_{BS} Supply Under−Voltage Negative−Going Threshold	$V_{BS1,2,3}$ = Sweep	7.0	8.0	8.7	V
V_{BSHYS}	V_{BS} Supply Under−Voltage Lockout Hysteresis	$V_{BS1,2,3}$ = Sweep	−	0.5	−	V
I_{LK}	Offset Supply Leakage Current	$V_{B1,2,3}$ = $V_{S1,2,3}$ = 600 V	−	−	10	μA
I_{QBS}	Quiescent V_{BS} Supply Current	$V_{HIN1,2,3}$ = 0 V or 5 V, EN = 0 V	10	50	80	μA
I_{PBS}	Operating V_{BS} Supply Current	$f_{HIN1,2,3}$ = 20 kHz, rms Value	200	320	480	μA
GATE DRIVER OUTPUT SECTION						
V_{OH}	High−Level Output voltage, V_{BIAS} − V_O	I_O = 0 mA (No Load)	−	−	100	mV
V_{OL}	Low−Level Output voltage, V_O	I_O = 0 mA (No Load)	−	−	100	mV
I_{O+}	Output HIGH Short−Circuit Pulse Current (Note 4)	V_O = 0 V, V_{IN} = 0 V with PW ≤ 10 μs	250	350	−	mA
$I_{O−}$	Output LOW Short−Circuit Pulsed Current (Note 4)	V_O = 15 V, V_{IN} = 5 V with PW ≤ 10 μs	500	650	−	mA
V_S	Allowable Negative V_S Pin Voltage for HIN Signal Propagation to HO		−	−9.8	−9.0	V
LOGIC INPUT SECTION						
V_{IH}	Logic “0” Input Voltage $\overline{HIN1,2,3}$, $\overline{LIN1,2,3}$		2.5	−	−	V
V_{IL}	Logic “1” Input Voltage $\overline{HIN1,2,3}$, $\overline{LIN1,2,3}$		−	−	0.8	V
I_{IN+}	Logic Input Bias Current (HO = LO = HIGH)	V_{IN} = 0 V	77	100	143	μA
$I_{IN−}$	Logic Input Bias Current (HO = LO = LOW)	V_{IN} = 5 V	−	8.5	25	μA
R_{IN}	Logic Input Pull−Up Resistance		35	50	65	kΩ
ENABLE CONTROL SECTION (EN)						
V_{EN+}	Enable Positive−Going Threshold Voltage		2.5	−	−	V
$V_{EN−}$	Enable Negative−Going Threshold Voltage		−	−	0.8	V
I_{EN+}	Logic Enable “1” Input Bias Current	V_{EN} = 5 V (Pull−Down = 150 kΩ)	15	33	50	μA
$I_{EN−}$	Logic Enable “0” Input Bias Current	V_{EN} = 0 V	−	−	2	μA
R_{EN}	Logic Input Pull−Down Resistance		100	150	333	kΩ
OVER−CURRENT PROTECTION SECTION						
V_{CSTH+}	Over−Current Detect Positive Threshold		450	500	550	mV
$V_{CSTH−}$	Over−Current Detect Negative Threshold		−	440	−	mV
V_{CSHYS}	Over−Current Detect Hysteresis		−	60	−	mV
I_{CSIN}	Short−Circuit Input Current	V_{CSIN} = 1 V	5	10	15	μA
I_{SOFT}	Soft Turn−Off Sink Current		25	40	55	mA

ELECTRICAL CHARACTERISTICS (V_{BIAS} (V_{DD}, $V_{BS1,2,3}$) = 15.0 V and T_A = 25°C unless otherwise specified. The V_{IN} and I_{IN} parameters are referenced to V_{SS} and are applicable to all six channels. The V_O and I_O parameters are referenced to $V_{S1,2,3}$ and COM and are applicable to the respective output leads: HO1,2,3 and LO1,2,3. The V_{DDUV} parameters are referenced to V_{SS}. The V_{BSUV} parameters are referenced to $V_{S1,2,3}$.) (continued)

Symbol	Parameter	Condition	Min	Typ	Max	Unit
FAULT OUTPUT SECTION						
$V_{RCINTH+}$	RCIN Positive−Going Threshold Voltage		2.7	3.3	3.9	V
$V_{RCINTH-}$	RCIN Negative−Going Threshold Voltage (Note 4)		−	2.6	−	V
$V_{RCINHYS}$	RCIN Hysteresis Voltage (Note 4)		−	0.7	−	V
I_{RCIN}	RCIN Internal Current Source	C_{RCIN} = 2 nF	3	5	7	μA
V_{FOL}	Fault Output Low Level Voltage	V_{CS} = 1 V, I_{FO} = 1.5 mA	−	0.2	0.5	V
R_{DSRCIN}	RCIN On Resistance	I_{RCIN} = 1.5 mA	50	75	100	Ω
R_{DSFO}	Fault Output On Resistance	I_{FO} = 1.5 mA	90	130	170	Ω

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

4. These parameters are guaranteed by design.

DYNAMIC ELECTRICAL CHARACTERISTICS (T_A = 25°C, V_{BIAS} (V_{DD}, $V_{BS1,2,3}$) = 15.0 V, $V_{S1,2,3}$ = COM = V_{SS}, C_{RCIN} = 2 nF, and C_{Load} = 1000 pF unless otherwise specified.)

Symbol	Parameter	Condition	Min	Typ	Max	Unit
t_{ON}	Turn−On Propagation Delay	$V_{LIN1,2,3}$ = $V_{HIN1,2,3}$ = 0 V, $V_{S1,2,3}$ = 0 V	350	500	650	ns
t_{OFF}	Turn−Off Propagation Delay	$V_{LIN1,2,3}$ = $V_{HIN1,2,3}$ = 5 V, $V_{S1,2,3}$ = 0 V	350	500	650	ns
t_R	Turn−On Rise Time	$V_{LIN1,2,3}$ = $V_{HIN1,2,3}$ = 0 V	20	50	100	ns
t_F	Turn−Off Fall Time	$V_{LIN1,2,3}$ = $V_{HIN1,2,3}$ = 5 V	10	30	80	ns
t_{EN}	Enable LOW to Output Shutdown Delay		400	500	600	ns
t_{CSBLT}	CS Pin Leading−Edge Blanking Time		400	650	850	ns
t_{CSFO}	Time from CS Triggering to $\overline{FO}$	From V_{CSC} = 1 V to $\overline{FO}$ Turn−Off	−	850	1300	ns
t_{CSOFF}	Time from CS Triggering to Low−Side Gate Outputs Turn−Off	From V_{CSC} = 1 V to Starting Gate Turn−Off	−	850	1300	ns
t_{FLTIN}	Input Filtering Time (Note 5) ($\overline{HINx}$, $\overline{LINx}$, EN)		170	250	330	ns
t_{FLTCLR}	Fault−Clear Time	C_{RCIN} = 2 nF	−	1.30	2.35	ms
DT	Dead Time		230	320	400	ns
MDT	Dead−Time Matching (All Six Channels)		−	−	50	ns
MT	Delay Matching (All Six Channels)		−	−	50	ns
PM	Output Pulse−Width Matching (Note 6)	PW_{IN} > 1 μs	−	50	100	ns

5. The minimum width of the input pulse should exceed 500 ns to ensure the filtering time of the input filter is exceeded.
6. PM is defined as $PW_{IN} - PW_{OUT}$.

TYPICAL CHARACTERISTICS

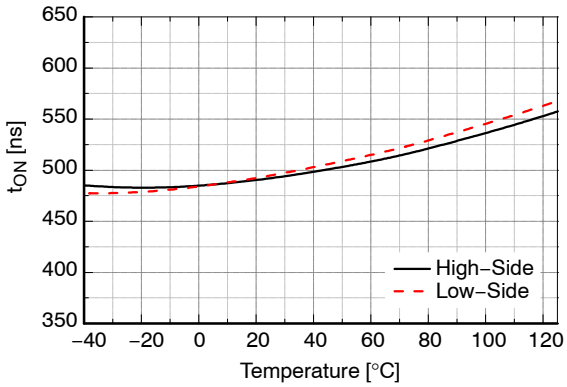


Figure 4. Turn−On Propagation Delay vs. Temperature

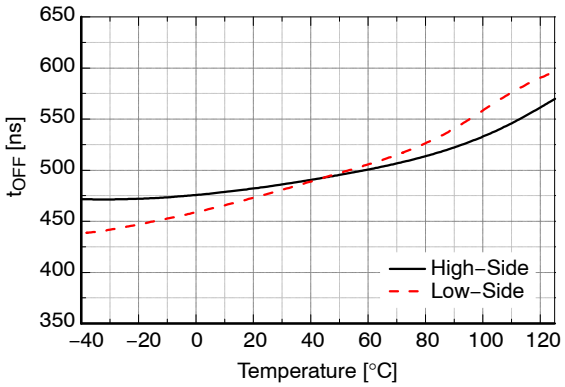


Figure 5. Turn−Off Propagation Delay vs. Temperature

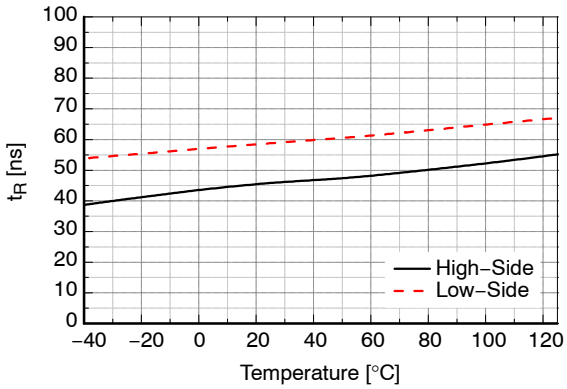


Figure 6. Turn−On Rise Time vs. Temperature

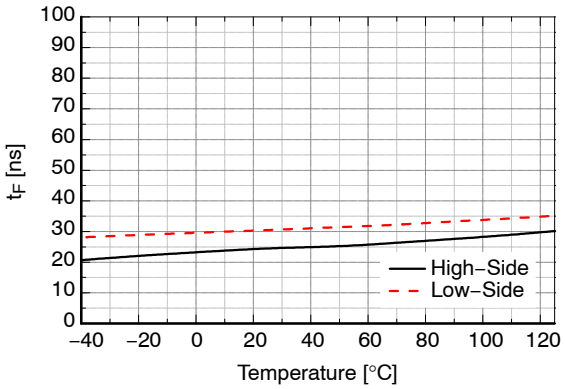


Figure 7. Turn−Off Fall Time vs. Temperature

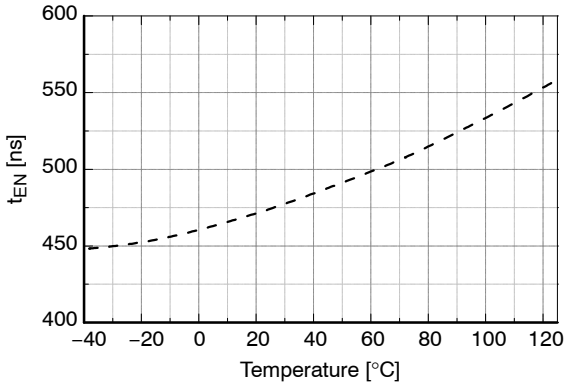


Figure 8. Enable LOW to Output Shutdown Delay vs. Temperature

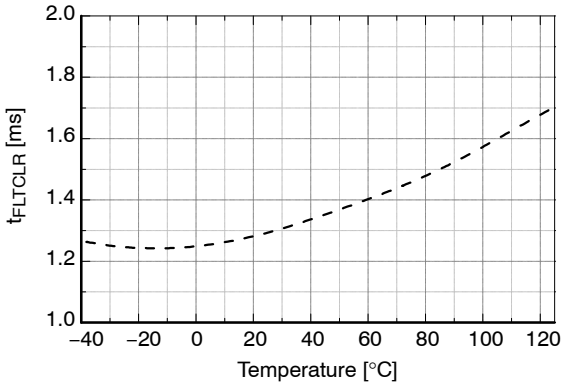


Figure 9. Fault−Clear Time vs. Temperature

TYPICAL CHARACTERISTICS (continued)

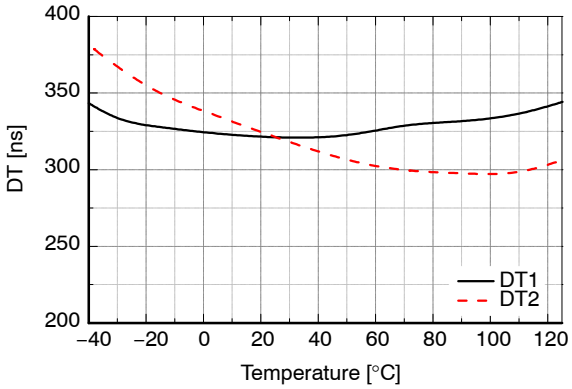


Figure 10. Dead Time vs. Temperature

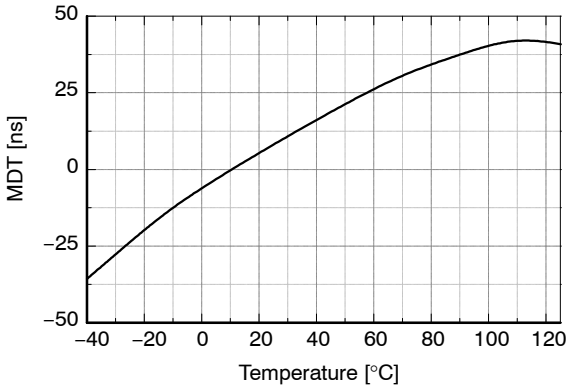


Figure 11. Dead−Time Matching vs. Temperature

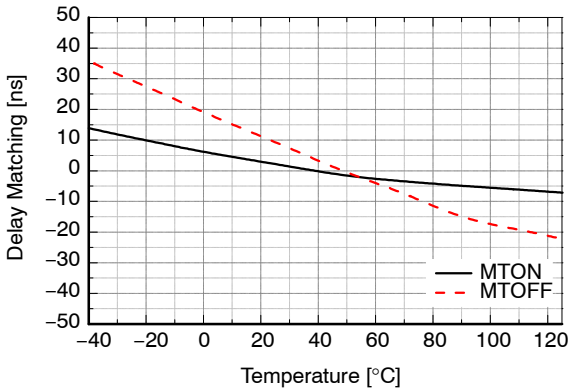


Figure 12. Delay Matching vs. Temperature

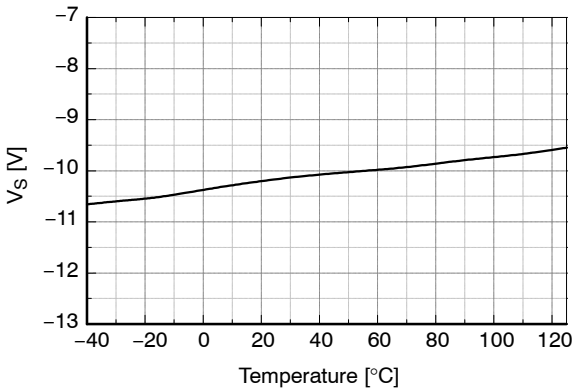


Figure 13. Allowable Negative V_S Voltage vs. Temperature

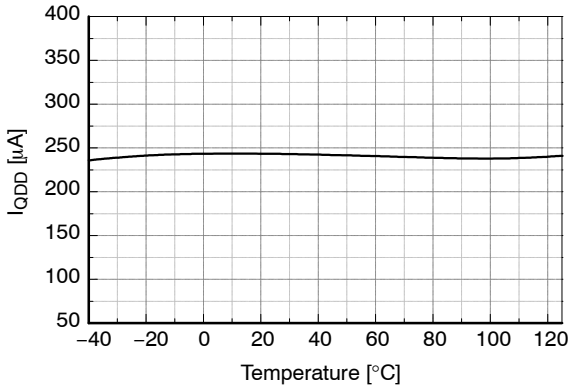


Figure 14. Quiescent V_{DD} Supply Current vs. Temperature

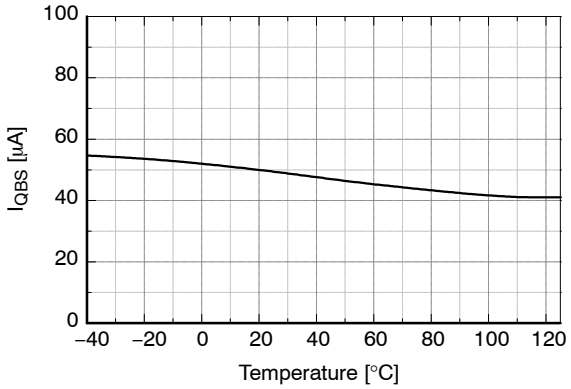


Figure 15. Quiescent V_{BS} Supply Current vs. Temperature

TYPICAL CHARACTERISTICS (continued)

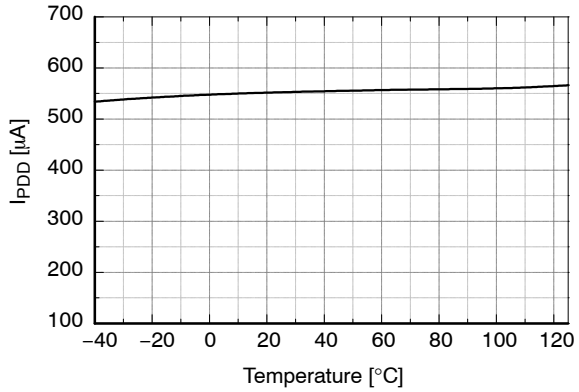


Figure 16. Operating V_{DD} Supply Current vs. Temperature

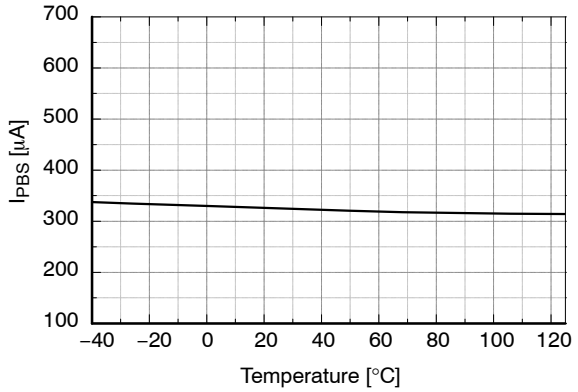


Figure 17. Operating V_{BS} Supply Current vs. Temperature

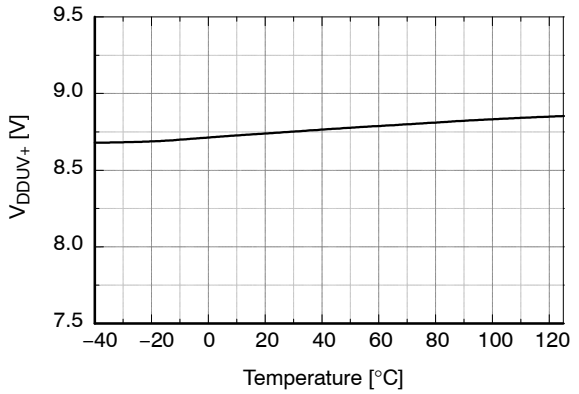


Figure 18. V_{DD} UVLO+ vs. Temperature

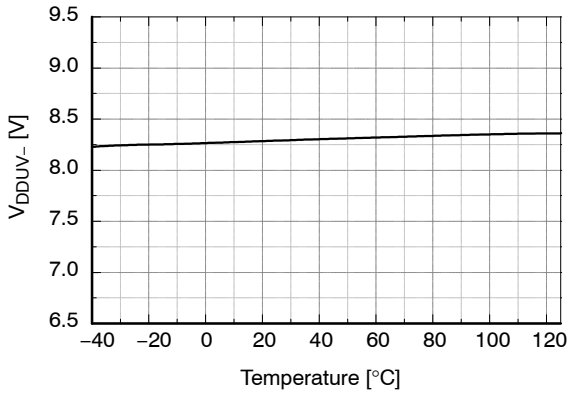


Figure 19. V_{DD} UVLO− vs. Temperature

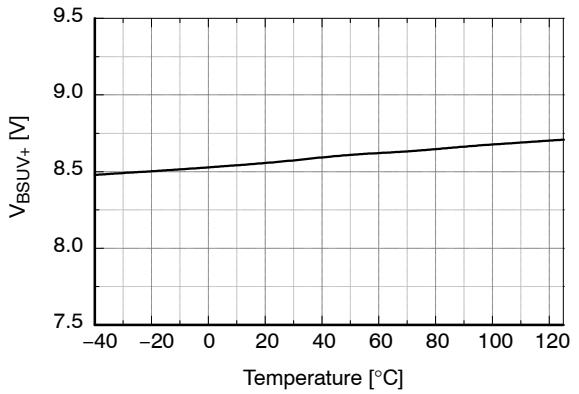


Figure 20. V_{BS} UVLO+ vs. Temperature

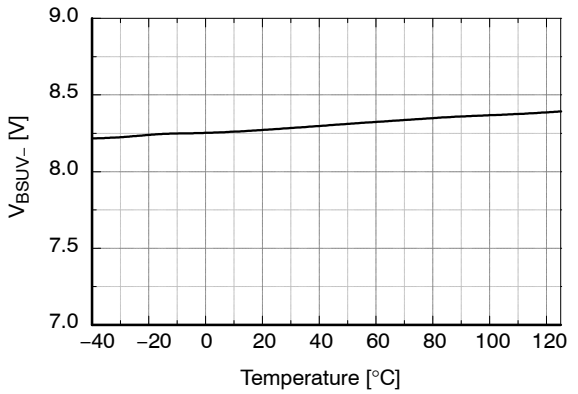


Figure 21. V_{BS} UVLO− vs. Temperature

TYPICAL CHARACTERISTICS (continued)

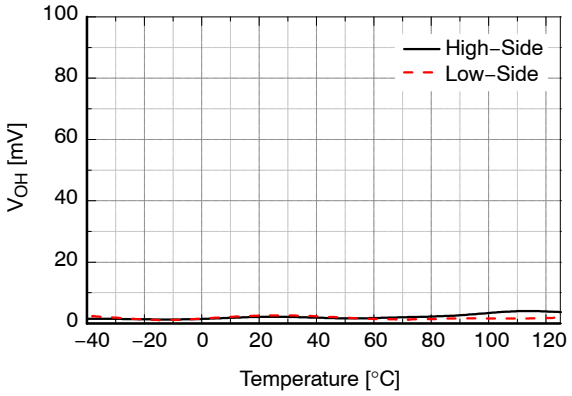


Figure 22. High−Level Output Voltage vs. Temperature

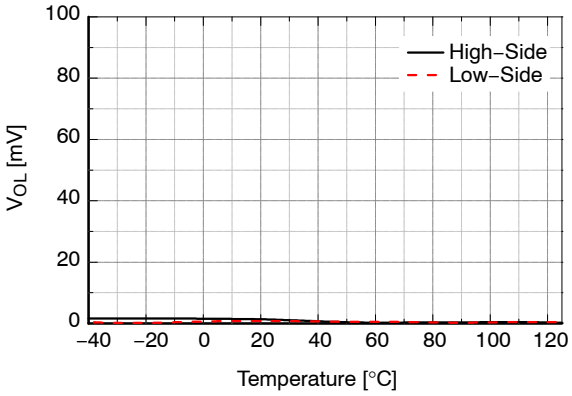


Figure 23. Low−Level Output Voltage vs. Temperature

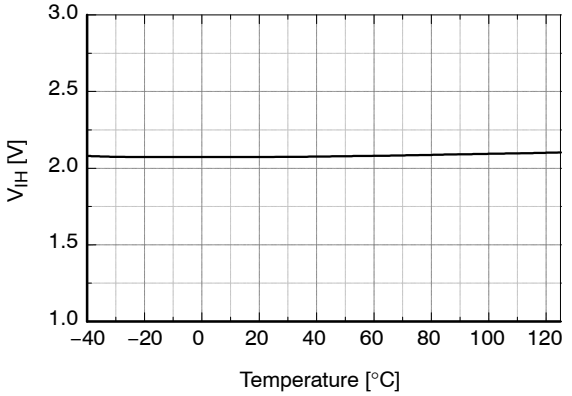


Figure 24. Logic HIGH Input Voltage vs. Temperature

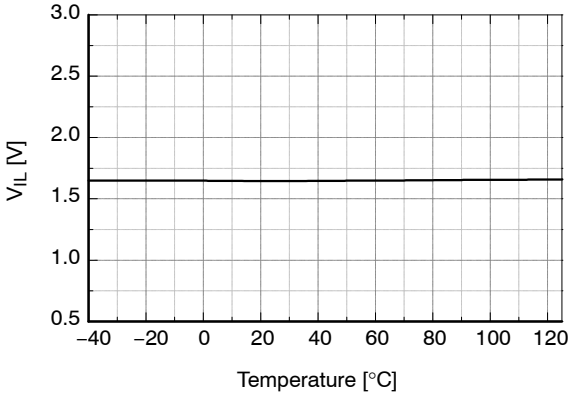


Figure 25. Logic LOW Input Voltage vs. Temperature

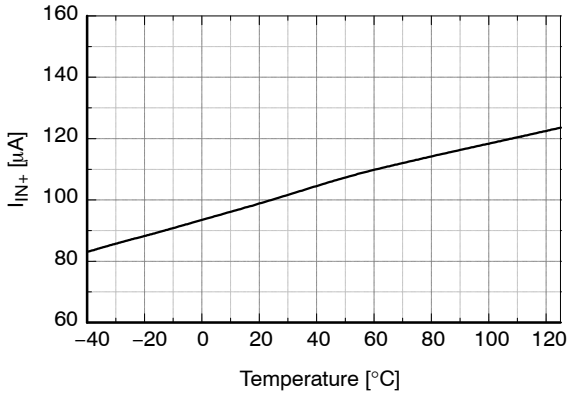


Figure 26. Logic Input HIGH Bias Current vs. Temperature

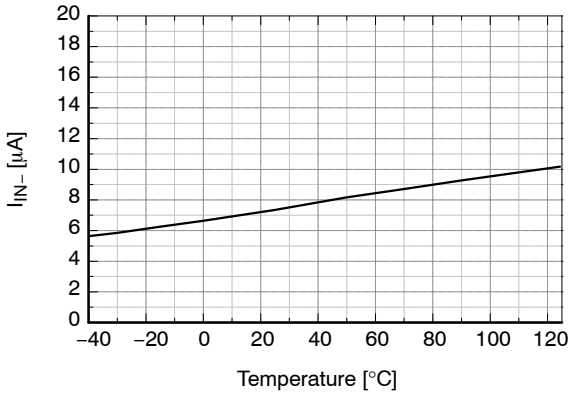


Figure 27. Logic Input LOW Bias Current vs. Temperature

TYPICAL CHARACTERISTICS (continued)

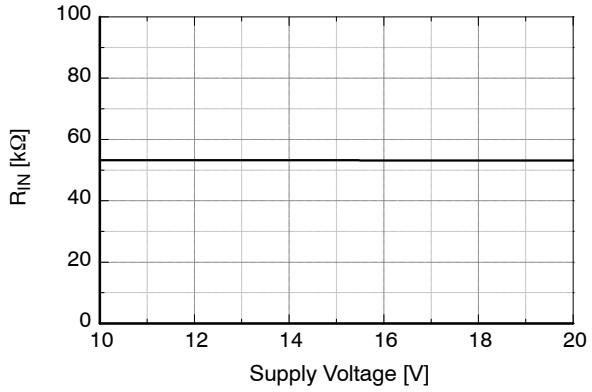


Figure 28. Input Pull−Down Resistance vs. Supply Voltage

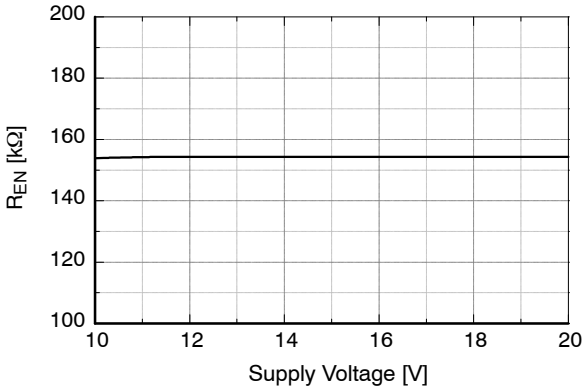


Figure 29. Enable Pin Pull−Down Resistance vs. Supply Voltage

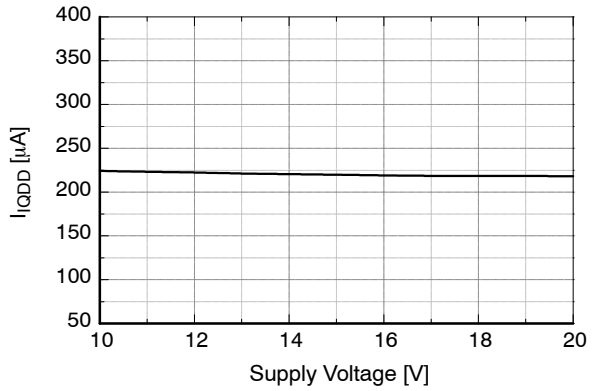


Figure 30. Quiescent V_{DD} Supply Current vs. Supply Voltage

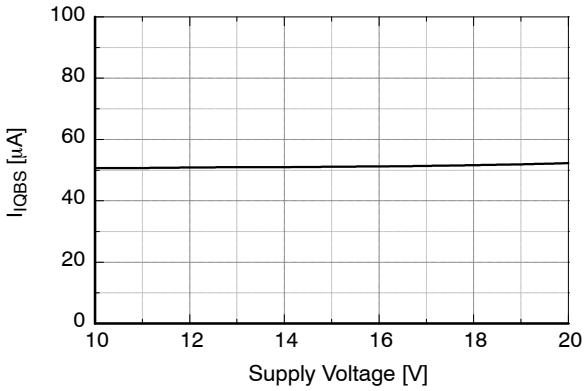


Figure 31. Quiescent V_{BS} Supply Current vs. Supply Voltage

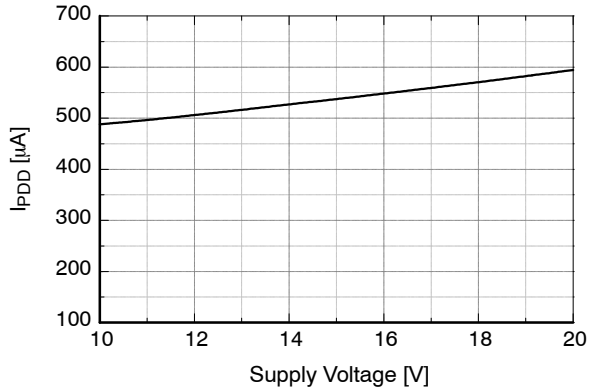


Figure 32. Operating V_{DD} Supply Current vs. Supply Voltage

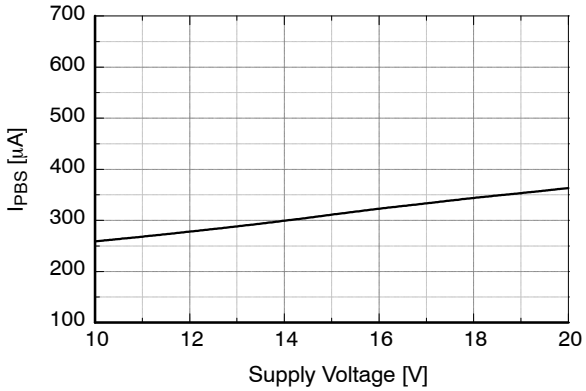


Figure 33. Operating V_{BS} Supply Current vs. Supply Voltage

SWITCHING TIME DEFINITIONS

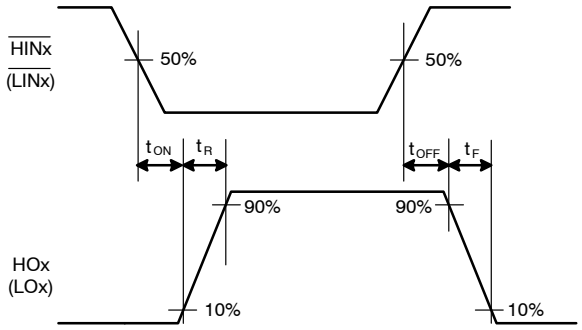


Figure 34. Switching Time Waveform Definitions

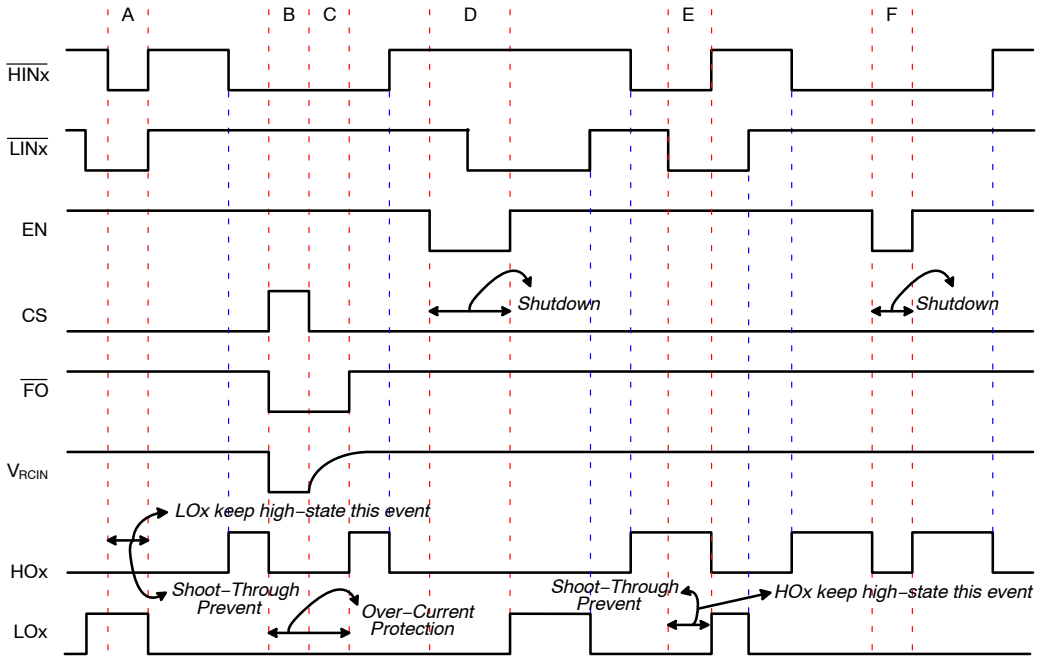


Figure 35. Input / Output Timing Diagram

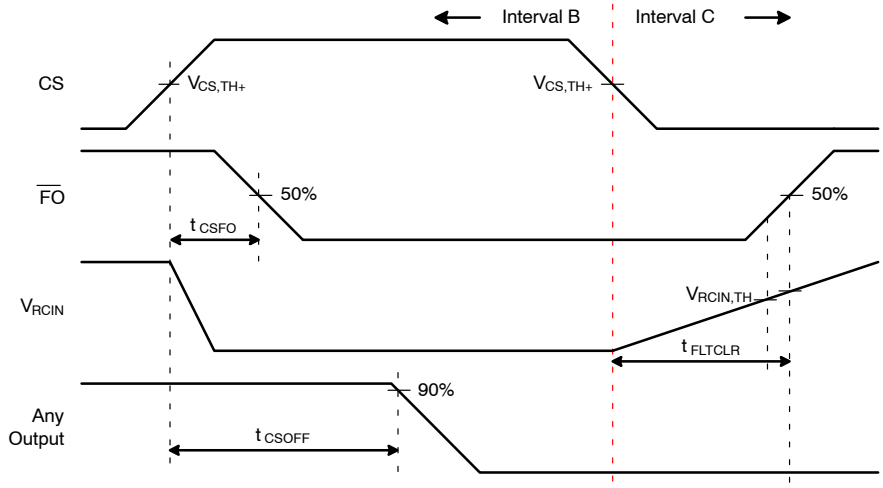


Figure 36. Detailed View of B and C Intervals During Over−Current Protection

APPLICATIONS INFORMATION

Dead Time

Dead time is automatically inserted whenever the dead time of the external two input signals (between $\overline{\text{HINx}}$ and $\overline{\text{LINx}}$ signals) is shorter than internal fixed dead times (DT1 and DT2). Otherwise, external dead times larger than internal dead times are not modified by the gate driver and internal dead−time waveform definition is shown in Figure 37.

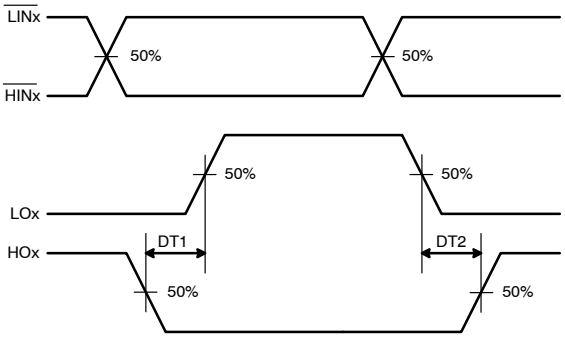


Figure 37. Internal Dead−Time Definitions

Protection Function

Fault Out ($\overline{FO}$) and Under−Voltage Lockout

The high− and low−side drivers include Under−Voltage Lockout (UVLO) protection circuitry that monitors the supply voltage for V_{DD} and V_{BS} independently. It can be designed to prevent malfunction when V_{DD} and V_{BS} are lower than the specified threshold voltage. The UVLO hysteresis prevents chattering during power−supply transitions. Moreover, the fault signal (power supply voltage $\overline{\text{FO}}$) goes to LOW state to operate reliably during power−on events when the power supply (V_{DD}) is below the under−voltage lockout high threshold voltage for the circuit (during t_1~t_2). The UVLO circuit is not otherwise activated; shown Figure 38. If V_{DD} is lower than 3.5 V, the fault signal cannot be driven to LOW state because V_{DD} is not enough to drive internal circuit.

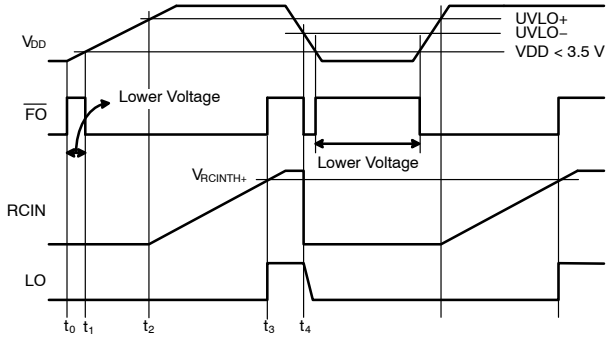


Figure 38. Waveforms for Under−Voltage Lockout

Shoot−Through Protection

The shoot−through protection circuitry prevents both high− and low−side switches from conducting at the same time, as shown Figure 39.

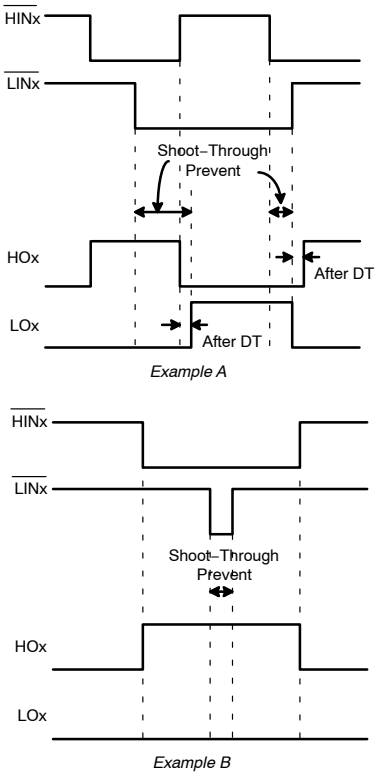
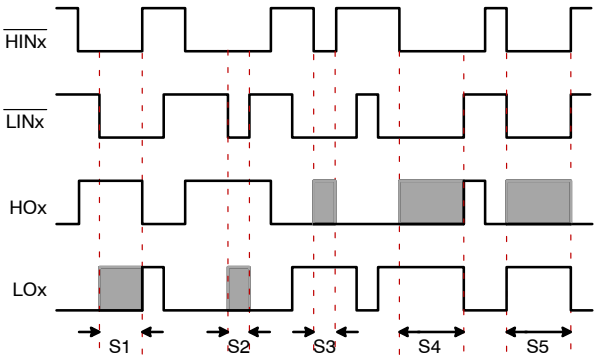


Figure 39. Shoot−Through Protection

An interlock function is a device used to prevent both high− and low−side switches from conducting at the same time as shown Figure 40. In most applications an interlock is used to help prevent a device from harming its operator or damaging itself by when two input signals of a same leg are activated simultaneously, only one output is activated.



S1 : High−side first →First input output mode
S2 : Low−side noise →No LOx output mode
S3 : High−side noise→No HOx output mode
S4 : Low−side first →First input output mode
S5 : In−phase mode →No HOx output

Figure 40. Interlock Function

Enable Input

When the EN pin is in HIGH state, the gate driver operates normally. When a condition occurs that should shut down the gate driver, the EN pin should be LOW. The enable circuitry has an input filter; the minimum input duration is specified by t_{FLTIN} (typically 250 ns).

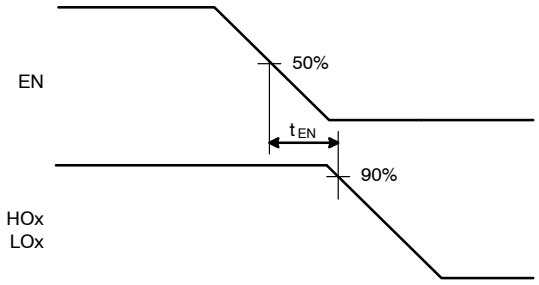


Figure 41. Output Enable Timing Waveform

Fault−Out ($\overline{FO}$) and Over−Current Protection

FAN73893 provides an integrated fault output ($\overline{FO}$) and an adjustable fault−clear timer (t_{FLTCLR}). There are two situations that cause the gate driver to report a fault via the $\overline{FO}$ pin. The first is an under−voltage condition of low−side gate driver supply voltage (V_{DD}) and the second is when the current−sense pin (CS) recognizes a fault. If a fault condition occurs, the $\overline{FO}$ pin is internally pulled to COM, the fault−clear timer is activated, and all outputs (HO1, 2, 3 and LO1, 2, 3) of the gate driver are turned off. The fault output stays LOW until the fault condition has been removed and the fault−clear timer expires. Once the fault−clear timer expires, the voltage on the $\overline{FO}$ pin returns to pull−up voltage.

The fault−clear time (t_{FLTCLR}) is determined by an internal current source (I_{RCIN} = 5 μA) and an external C_{RCIN} at the RCIN pin, as shown as:

$$t_{FLTCLR} = \frac{C_{RCIN} \times V_{RCIN,TH}}{I_{RCIN}} \text{ [s]} \qquad \text{(eq. 1)}$$

The R_{DSRCIN} of the MOSFET is a characteristic discharge curve with respect to the external capacitor C_{RCIN}. The time constant is defined by the external capacitor C_{RCIN} and the R_{DSRCIN} of the MOSFET.

The output of current−sense comparator (CS_COMP) passes a noise filter, which inhibits an over−current shutdown caused by parasitic voltage spikes of V_{CS}.

This corresponds to a voltage level at the comparator of $V_{CSTH+} - V_{CSHYS}$ = 500 mV − 60 mV = 440 mV, where V_{CSHYS} = 60 mV is the hysteresis of the current comparator (CS_COMP), as shown in Figure 42.

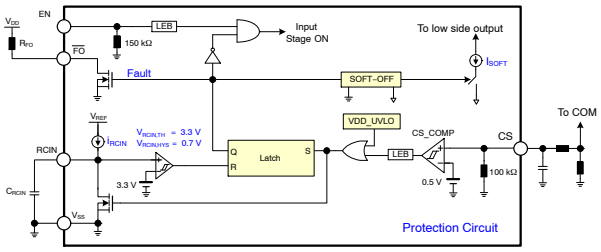


Figure 42. Over−Current Protection

Figure 43 shows the waveform definitions of RCIN, $\overline{FO}$, and the low−side driver; which uses a soft turn−off method when an under−voltage condition of the low−side gate driver supply voltage (V_{DD}) or the current−sense pin (CS) recognizes a fault. If a fault condition occurs, the $\overline{FO}$ Pin is internally pulled to COM and all outputs (HO1,2,3 and LO1,2,3) of the gate driver are turned off. Low−side outputs decline linearly by the internal sink current source (I_{SOFT} = 40 mA) for soft turn−off, as shown in Figure 43.

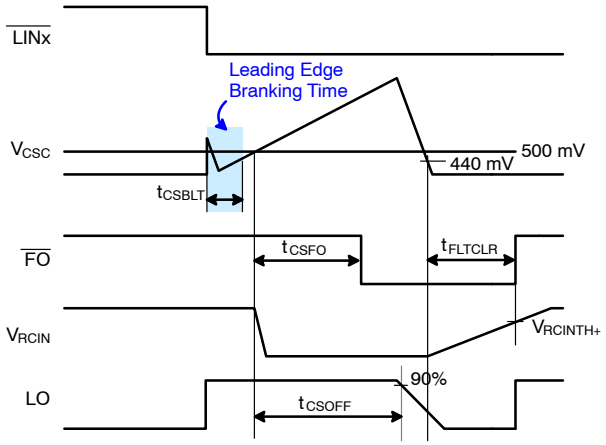


Figure 43. R_{CIN} and Fault−Clear Waveform Definition

Noise Filter

Input Noise Filter

Figure 44 shows the input noise filter method, which has symmetry duration between the input signal (t_{INPUT}) and the output signal (t_{OUTPUT}) and helps to reject noise spikes and short pulses. This input filter is applied to the HINx, LINx, and EN inputs. The upper pair of waveforms (Example A) shows an input signal duration (t_{INPUT}) much longer than input filter time (t_{FLTIN}); it is approximately the same duration between the input signal time (t_{INPUT}) and the output signal time (t_{OUTPUT}). The lower pair of waveforms (Example B) shows an input signal time (t_{INPUT}) slightly longer than input filter time (t_{FLTIN}); it is approximately the same duration between input signal time (t_{INPUT}) and the output signal time (t_{OUTPUT}).

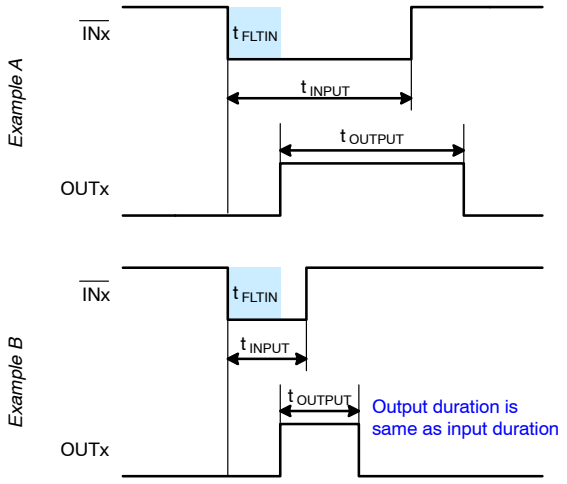


Figure 44. Input Noise Filter Definition

Short−Pulsed Input Noise Rejection Method

The input filter circuitry provides protection against short−pulsed input signals ($\overline{\text{HINx}}$, $\overline{\text{LINx}}$ and EN) on the input signal lines by applied noise signal.

If the input signal duration is less than input filter time (t_{FLTIN}), the output does not change states.

Example A and B of the Figure 45 show the input and output waveforms with short−pulsed noise spikes with a duration less than input filter time; the output does not change states.

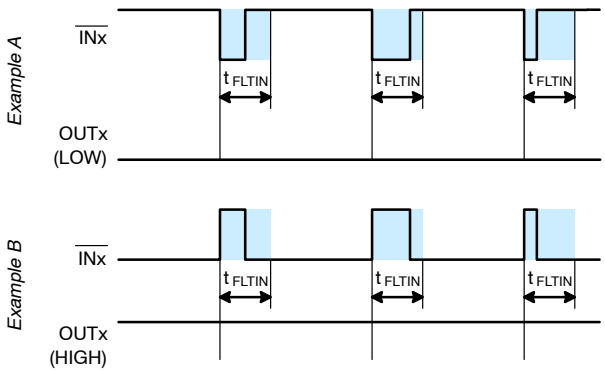


Figure 45. Noise Rejecting Input Filter Definition

Figure 46 shows the characteristics of the input filters while receiving narrow ON and OFF pulses. If input signal pulse duration, PW_{IN}, is less than input filter time, t_{FLTIN}; the output pulse, PW_{OUT}, is zero. The input signal is rejected by input filter. Once the input signal pulse duration, PW_{IN}, exceeds input filter time, t_{FLTIN}, the output pulse durations, PW_{OUT}, matches the input pulse durations, PW_{IN}. FAN73893 input filter time, t_{FLTIN}, is about 250 ns for the high− and low−side outputs.

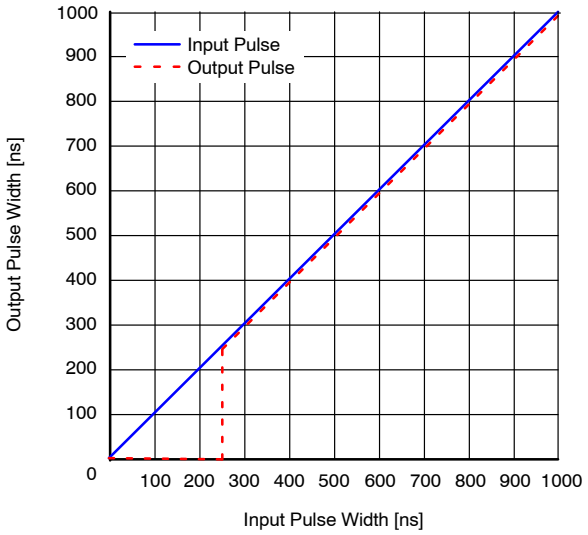


Figure 46. Input Filter Characteristic of Narrow ON

ORDERING INFORMATION

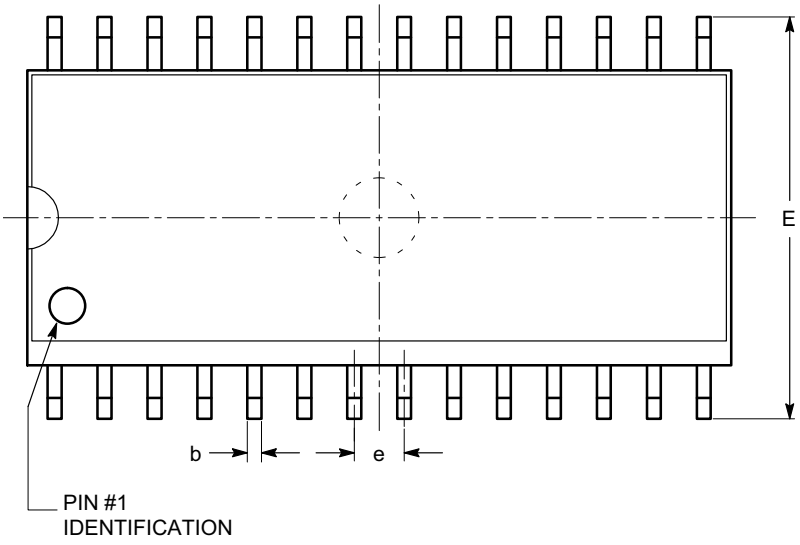
Part Number	Package	Operating Temperature	Shipping†
FAN73893MX (Note 7)	28−Lead, Small Outline Integrated Circuit, (SOIC) (Pb−Free)	−40 to +125°C	1000 / Tape & Reel

†For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

7. These devices passed wave−soldering test by JESD22A−111.

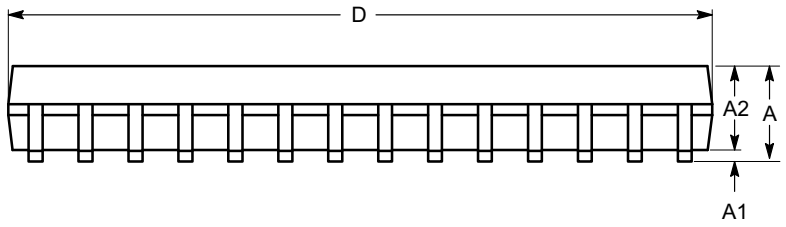
SOIC−28, 300 mils
CASE 751BM
ISSUE O

DATE 19 DEC 2008

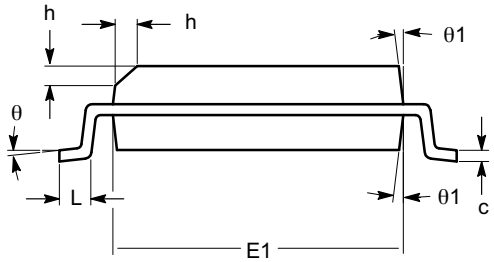


TOP VIEW

SYMBOL	MIN	NOM	MAX
A	2.35		2.65
A1	0.10		0.30
A2	2.05		2.55
b	0.31		0.51
c	0.20		0.33
D	17.78		18.03
E	10.11		10.51
E1	7.34		7.60
e		1.27 BSC	
h	0.25		0.75
L	0.40		1.27
θ	0º		8º
θ1	5º		15º



SIDE VIEW



END VIEW

Notes:
(1) All dimensions are in millimeters. Angles in degrees.
(2) Complies with JEDEC MS-013.

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DESCRIPTION:	SOIC−28, 300 MILS	PAGE 1 OF 1

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